# Random Matrix Theory for Signal Processing Applications

Romain Couillet<sup>1</sup>, Mérouane Debbah<sup>2</sup>

<sup>1</sup>EDF Chair on System Sciences and the Energy Challenge, Supélec, Gif sur Yvette, France <sup>2</sup>Alcatel-Lucent Chair on Flexible Radio, Supélec, Gif sur Yvette, FRANCE {romain.couillet,merouane.debbah}@supelec.fr

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### Tools for Random Matrix Theory

- Classical Random Matrix Theory
- Introduction to Large Dimensional Random Matrix Theory
- The Random Matrix Pioneers
- The Moment Approach and Free Probability
- Introduction of the Stieltjes Transform
- Properties of the Asymptotic Support and Spiked Models
- Summary of what we know and what is left to be done

### 2 Random Matrix Theory and Signal Source Sensing

- Small Dimensional Analysis
- Large Dimensional Random Matrix Analysis

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- Optimal detector
- The moment method
- The Stieltjes transform method

### 4 Random Matrix Theory and Failure Detection in Complex Systems

- Random matrix models of local failures in sensor networks
- Failure detection and localization

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# Definitions

#### Random Matrix

A random matrix is a matrix  $\mathbf{X} \in \mathbb{C}^{N \times n}$  with random entries  $X_{ij}$  following a given probability distribution.

- In many problems (with symmetrical structures), interest is on:
  - eigenvalue distribution
  - eigenvector projections.

Pioneering works due to Wishart on matrices

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with  $X_{ij} \sim \mathcal{CN}(0,1)$ 

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# Wishart matrices

J. Wishart, "The generalized product moment distribution in samples from a normal multivariate population", Biometrika, vol. 20A, pp. 32-52, 1928.

• Wishart describes the distribution of  $\mathbf{R}_n = \mathbf{X}\mathbf{X}^H = \sum_{i=1}^n \mathbf{x}_i \mathbf{x}_i^H$ ,  $\mathbf{x}_i \in \mathbb{C}^N \sim \mathcal{CN}(0, \mathbf{R})$ ,

$$P_{\mathbf{R}_n}(\mathbf{B}) = \frac{\pi^{N(N-1)/2}}{\det \mathbf{R}^n \prod_{i=1}^N (n-i)!} e^{-\operatorname{tr}(\mathbf{R}^{-1}\mathbf{B})} \det \mathbf{B}^{n-N}$$

• Joint and marginal eigenvalue distributions:

$$P_{(\lambda_i)}(\lambda_1,\ldots,\lambda_N) = \frac{\det(\{e^{-r_j^{-1}\lambda_i}\}_N)}{\Delta(\mathbf{R}^{-1})}\Delta(\mathbf{L})\prod_{j=1}^N \frac{\lambda_j^{n-N}}{j!(n-j)!}$$

with  $r_1 \ge \ldots \ge r_N$  the eigenvalues of **R** and **L** = diag( $\lambda_1 \ge \ldots \ge \lambda_N$ ) and

$$p_{\lambda}(\lambda) = \frac{1}{M} \sum_{k=0}^{N-1} \frac{k!}{(k+n-N)!} [L_k^{n-N}]^2 \lambda^{n-N} e^{-\lambda}$$

where  $L_n^k$  are the Laguerre polynomials

$$L_n^k(\lambda) = \frac{e^{\lambda}}{k!\lambda^n} \frac{d^k}{d\lambda^k} (e^{-\lambda} \lambda^{n+k}).$$

#### Extensions to:

- correlated Gaussian involve heavy tools (Schur polynomials)
- non-Gaussian is virtually impossible!
- Solution is to assume increasing matrix dimensions:  $N, n \rightarrow \infty$ 
  - deterministic limiting behaviour is often observed
  - loose assumptions on entry distributions (e.g. rotational symmetry, independent entries)
  - robust framework for very generic models are known:
    - Stieltjes transform methods (more efficient than Fourier transform)
    - moments/free probability methods (extension of classical probability for non-commutative variables)
    - physical methods for large systems (replica method)

This tutorial will introduce the major used methods but concentrates on the powerful Stieltjes transform method.

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Let  $\mathbf{w}_1, \mathbf{w}_2 \ldots \in \mathbb{C}^N$  be independently drawn from an *N*-variate process of mean zero and covariance  $\mathbf{R} = \mathrm{E}[\mathbf{w}_1\mathbf{w}_1^H] \in \mathbb{C}^{N \times N}$ .

#### Law of large numbers

As  $n \to \infty$ ,

$$\frac{1}{n}\sum_{i=1}^{n}\mathbf{w}_{i}\mathbf{w}_{i}^{\mathsf{H}}=\mathbf{W}\mathbf{W}^{\mathsf{H}}\xrightarrow{\text{a.s.}}\mathbf{R}$$

In reality, one cannot afford  $n \to \infty$ .

• if  $n \gg N$ ,

$$\mathbf{R}_n = \frac{1}{n} \sum_{i=1}^n \mathbf{w}_i \mathbf{w}_i^{\mathsf{H}}$$

is a "good" estimate of **R**.

• if N/n = O(1), and if both (n, N) are large, we can still say, for all (i, j),

$$(\mathbf{R}_n)_{ij} \stackrel{\mathrm{a.s.}}{\longrightarrow} (\mathbf{R})_{ij}$$

What about the global behaviour? What about the eigenvalue distribution?

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#### Tools for Random Matrix Theory Introduction to Large Dimensional Random Matrix Theor Empirical and limit spectra of Wishart matrices



Figure: Histogram of the eigenvalues of  $\mathbf{R}_n$  for n = 2000, N = 500,  $\mathbf{R} = \mathbf{I}_N$ 

# The Marčenko-Pastur Law



Figure: Marčenko-Pastur law for different limit ratios  $c = \lim N/n$ .

# The Marčenko-Pastur law

Let  $\mathbf{W} \in \mathbb{C}^{N \times n}$  have i.i.d. elements, of zero mean and variance 1/n. Eigenvalues of the matrix



when  $N, n \rightarrow \infty$  with  $N/n \rightarrow c$  **IS NOT IDENTITY!** 

**Remark:** If the entries are Gaussian, the matrix is called a Wishart matrix with *n* degrees of freedom. The **exact** distribution is known in the finite case.

Introduction to Large Dimensional Random Matrix Theory

# Deriving the Marčenko-Pastur law

• We wish to determine the density  $f_c(\lambda)$  of the asymptotic law, defined by

$$f_{c}(\lambda) = \lim_{\substack{N \to \infty \\ n \to \infty \\ N/n \to c}} \sum_{i=1}^{N} \delta\left(\lambda - \lambda_{i}(\mathbf{R}_{n})\right)$$

• With  $N/n \rightarrow c$ , the moments of this distribution are given by

$$M_1^N = \frac{1}{N} \operatorname{tr} \mathbf{R}_n = \frac{1}{N} \sum_{i=1}^N \lambda_i(\mathbf{R}_n) \to \int \lambda f_c(\lambda) d\lambda = 1$$
  

$$M_2^N = \frac{1}{N} \operatorname{tr} \mathbf{R}_n^2 = \frac{1}{N} \sum_{i=1}^N \lambda_i(\mathbf{R}_n)^2 \to \int \lambda^2 f_c(\lambda) d\lambda = 1 + c$$
  

$$M_3^N = \frac{1}{N} \operatorname{tr} \mathbf{R}_n^3 = \frac{1}{N} \sum_{i=1}^N \lambda_i(\mathbf{R}_n)^3 \to \int \lambda^3 f_c(\lambda) d\lambda = c^2 + 3c + 1$$
  

$$\cdots = \cdots$$

 These moments correspond to a *unique* distribution function (under mild assumptions), which has density the Marčenko-Pastur law

$$f(x) = (1 - \frac{1}{c})^+ \delta(x) + \frac{\sqrt{(x - a)^+ (b - x)^+}}{2\pi cx}, \text{ with } a = (1 - \sqrt{c})^2, b = (1 + \sqrt{c})^2.$$

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# Wigner and semi-circle law

Schrödinger's equation

$$H\Phi_i = E_i \Phi_i$$

where  $\Phi_i$  is the wave function,  $E_i$  is the energy level, *H* is the Hamiltonian.



Magnetic interactions between the spins of electrons

The birth of large dimensional random matrix theory The Random Matrix Theory



Eugene Paul Wigner, 1902-1995

Random Matrix Theory for Signal Processing Applicat

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E. Wigner, "Characteristic vectors of bordered matrices with infinite dimensions," The annals of mathematics, vol. 62, pp. 546-564, 1955.

$$\mathbf{X}_{N} = \frac{1}{\sqrt{N}} \begin{bmatrix} 0 & +1 & +1 & +1 & -1 & -1 & \cdots \\ +1 & 0 & -1 & +1 & +1 & +1 & \cdots \\ +1 & -1 & 0 & +1 & +1 & +1 & \cdots \\ +1 & +1 & +1 & 0 & +1 & +1 & \cdots \\ -1 & +1 & +1 & +1 & 0 & -1 & \cdots \\ -1 & +1 & +1 & +1 & -1 & 0 & \cdots \\ \vdots & \ddots \end{bmatrix}$$

As the matrix dimension increases, what can we say about the eigenvalues (energy levels)?

If X<sub>N</sub> ∈ C<sup>N×N</sup> is Hermitian with i.i.d. entries of mean 0, variance 1/N above the diagonal, then F<sup>X<sub>N</sub></sup> a.s. F where F has density f the semi-circle law

$$f(x) = \frac{1}{2\pi} \sqrt{(4-x^2)^+}$$

Shown from the method of moments

$$\lim_{N\to\infty}\frac{1}{N}\operatorname{tr} \mathbf{X}_N^{2k} = \frac{1}{k+1}C_k^{2k}$$

#### which are exactly the moments of f(x)!

• If  $X_N \in \mathbb{C}^{N \times N}$  has i.i.d. 0 mean, variance 1/N entries, then asymptotically its complex eigenvalues distribute uniformly on the complex unit circle.

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# Semi-circle law



Figure: Histogram of the eigenvalues of Wigner matrices and the semi-circle law, for N = 500

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# Circular law



Figure: Eigenvalues of  $X_N$  with i.i.d. standard Gaussian entries, for N = 500.

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- much study has surrounded the Marčenko-Pastur law, the Wigner semi-circle law etc.
- for practical purposes, we often need more general matrix models
  - products and sums of random matrices
  - i.i.d. models with correlation/variance profile
  - distribution of inverses etc.
- for these models, it is often impossible to have a closed-form expression of the limiting distribution.
- sometimes we do not have a limiting convergence.

To study these models, the method of moments is not enough! A consistent powerful mathematical framework is required.

Image: A matrix

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• The Hermitian matrix  $\mathbf{R}_N \in \mathbb{C}^{N \times N}$  has successive *empirical* moments  $M_k^N$ , k = 1, 2, ...,

$$M_k^N = \frac{1}{N} \sum_{i=1}^N \lambda_i^k$$

• In classical probability theory, for A, B independent,

$$c_k(A+B) = c_k(A) + c_k(B)$$

with  $c_k(X)$  the cumulants of X. The cumulants  $c_k$  are connected to the moments  $m_k$  by,

$$m_k = \sum_{\pi \in \mathcal{P}(k)} \prod_{V \in \pi} c_{|V|}$$

A natural extension of classical probability for non-commutative random variables exist, called Free Probability

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A natural extension of classical probability for non-commutative random variables exist, called Free Probability

• The Hermitian matrix  $\mathbf{R}_N \in \mathbb{C}^{N \times N}$  has successive *empirical* moments  $M_k^N$ , k = 1, 2, ...,

$$M_k^N = \frac{1}{N} \sum_{i=1}^N \lambda_i^k$$

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### Free probability

*Free probability applies to asymptotically large random matrices. We denote the moments without superscript.* 

- To connect the moments of A + B to those of A and B, independence is not enough. A and B must be asymptotically free,
  - two Gaussian matrices are free
  - · a Gaussian matrix and any deterministic matrix are free
  - unitary (Haar distributed) matrices are free
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• Similarly as in classical probability, we define free cumulants  $C_k$ ,

 $C_1 = M_1$   $C_2 = M_2 - M_1^2$  $C_3 = M_3 - 3M_1M_2 + 2M_1^2$ 

R. Speicher, "Combinatorial theory of the free product with amalgamation and operator-valued free probability theory," Mem. A.M.S., vol. 627, 1998.

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## Non-crossing partitions



Figure: Non-crossing partition  $\pi = \{\{1, 3, 4\}, \{2\}, \{5, 6, 7\}, \{8\}\}$  of *NC*(8).

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### Tools for Random Matrix Theory The Moment Approach and Free Probability Moments of sums and products of random matrices

### Combinatorial calculus of all moments

#### Theorem

For free random matrices A and B, we have the relationship,

$$C_k(\mathbf{A} + \mathbf{B}) = C_k(\mathbf{A}) + C_k(\mathbf{B})$$

$$M_n(\mathbf{AB}) = \sum_{(\pi_1, \pi_2) \in NC(n)} \prod_{\substack{V_1 \in \pi_1 \\ V_2 \in \pi_2}} C_{|V_1|}(\mathbf{A}) C_{|V_2|}(\mathbf{B})$$

in conjunction with free moment-cumulant formula, gives all moments of sum and product.

#### Theorem

If F is a compactly supported distribution function, then F is determined by its moments.

• In the absence of support compactness, some conditions (e.g. Carleman) have to be checked. This is in particular the case of Vandermonde matrices.

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### Free convolution

• In classical probability theory, for independent A, B,

$$\mu_{A+B}(x) = \mu_A(x) * \mu_B(x) \stackrel{\Delta}{=} \int \mu_A(t) \mu_B(x-t) dt$$

• In free probability, for free A, B, we use the notations

$$\mu_{\mathbf{A}+\mathbf{B}} = \mu_{\mathbf{A}} \boxplus \mu_{\mathbf{B}}, \ \mu_{\mathbf{A}} = \mu_{\mathbf{A}+\mathbf{B}} \boxminus \mu_{\mathbf{B}}, \ \mu_{\mathbf{A}\mathbf{B}} = \mu_{\mathbf{A}} \boxtimes \mu_{\mathbf{B}}, \ \mu_{\mathbf{A}} = \mu_{\mathbf{A}+\mathbf{B}} \boxtimes \mu_{\mathbf{B}}$$

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#### Theorem

Convolution of the information-plus-noise model Let  $\mathbf{W}_N \in \mathbb{C}^{N \times n}$  have i.i.d. Gaussian entries of mean 0 and variance 1,  $\mathbf{A}_N \in \mathbb{C}^{N \times n}$ , such that  $\mu_{\frac{1}{n}\mathbf{A}_N\mathbf{A}_N^H} \Rightarrow \mu_A$ , as  $n/N \to c$ . Then the eigenvalue distribution of

$$\mathbf{B}_{N} = \frac{1}{n} \left( \mathbf{A}_{N} + \sigma \mathbf{W}_{N} \right) \left( \mathbf{A}_{N} + \sigma \mathbf{W}_{N} \right)^{\mathsf{H}}$$

converges weakly and almost surely to  $\mu_B$  such that

$$\mu_{B} = \left( (\mu_{A} \boxtimes \mu_{c}) \boxplus \delta_{\sigma^{2}} \right) \boxtimes \mu_{c}$$

with  $\mu_c$  the Marčenko-Pastur law with ratio c.

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Tools for Random Matrix Theory The Moment Approach and Free Probability Similarities between classical and free probability

	Classical Probability	Free probability
Moments	$m_k = \int x^k dF(x)$	$M_k = \int x^k dF(x)$
Cumulants	$m_n = \sum_{\pi \in \mathcal{D}(n)}^{s} \prod_{V \in \pi} c_{ V }$	$M_n = \sum_{\pi \in \mathcal{N} \cap (n)}^{s} \prod_{V \in \pi} C_{ V }$
Independence	$\pi \in \mathcal{P}(n)$ $V \in \pi$	freeness
Additive convolution	$f_{A+B} = f_A * f_B$	$\mu_{\mathbf{A}+\mathbf{B}} = \mu_{\mathbf{A}} \boxplus \mu_{\mathbf{B}}$
Multiplicative convolution	f <sub>AB</sub>	$\mu_{AB}=\mu_{A}\boxtimes\mu_{B}$
Sum Rule	$c_k(A+B) = c_k(A) + c_k(B)$	$C_k(\mathbf{A} + \mathbf{B}) = C_k(\mathbf{A}) + C_k(\mathbf{B})$
Central Limit	$\frac{1}{\sqrt{n}}\sum_{i=1}^n x_i \to \mathcal{N}(0,1)$	$\frac{1}{\sqrt{n}}\sum_{i=1}^{n}X_{i} \Rightarrow \text{semi-circle law}$

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### Tools for Random Matrix Theory The Moment Approach and Free Probability Bibliography on Free Probability related work

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### Outline

### Tools for Random Matrix Theory

- Classical Random Matrix Theory
- Introduction to Large Dimensional Random Matrix Theory
- The Random Matrix Pioneers
- The Moment Approach and Free Probability
- Introduction of the Stieltjes Transform
- Properties of the Asymptotic Support and Spiked Models
- Summary of what we know and what is left to be done

### 2 Random Matrix Theory and Signal Source Sensing

- Small Dimensional Analysis
- Large Dimensional Random Matrix Analysis

### 3 Random Matrix Theory and Multi-Source Power Estimation

- Optimal detector
- The moment method
- The Stieltjes transform method

### 4 Random Matrix Theory and Failure Detection in Complex Systems

- Random matrix models of local failures in sensor networks
- Failure detection and localization

## The Stieltjes transform

### Definition

Let *F* be a real distribution function. The Stieltjes transform  $m_F$  of *F* is the function defined, for  $z \in \mathbb{C} \setminus \mathbb{R}$ , as

$$m_F(z) = \int \frac{1}{\lambda - z} dF(\lambda)$$

For a < b real, denoting z = x + iy, we have the inverse formula

$$F'(x) = \lim_{y\to 0} \frac{1}{\pi} \Im[m_F(x+iy)]$$

Knowing the Stieltjes transform is knowing the eigenvalue distribution!

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## Remark on the Stieltjes transform

• If *F* is the eigenvalue distribution of a Hermitian matrix  $\mathbf{X}_N \in \mathbb{C}^{N \times N}$ , we might denote  $m_{\mathbf{X}} \triangleq m_F$ , and

$$m_{\mathbf{X}}(z) = \int \frac{1}{\lambda - z} dF(\lambda) = \frac{1}{N} \operatorname{tr} (\mathbf{X}_N - z \mathbf{I}_N)^{-1}$$

For compactly supported eigenvalue distribution,

$$m_F(z) = -\frac{1}{z} \int \frac{1}{1 - \frac{\lambda}{z}} = -\sum_{k=0}^{\infty} M_k^N z^{-k-1}$$

The Stieltjes transform is doubly more powerful than the moment approach!

- conveys more information than any *K*-finite sequence  $M_1, \ldots, M_K$ .
- is not handicapped by the support compactness constraint.
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- We wish to prove that the spectrum of  $XX^H$ ,  $X \in \mathbb{C}^{N \times n}$ , with entries  $\mathcal{CN}(0, 1/n)$  tends to the MP law.
- From a matrix inversion lemma

$$\left[ (\mathbf{X}\mathbf{X}^{\mathsf{H}} - z\mathbf{I}_{N})^{-1} \right]_{11} = \frac{1}{-z - z\mathbf{y}^{\mathsf{H}}(\mathbf{Y}^{\mathsf{H}}\mathbf{Y} - z\mathbf{I}_{n})^{-1}\mathbf{y}}$$

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Stieltjes transform proof of the Marčenko-Pastur law (2)

• This is a second order polynomial of the type

$$m_F(z) = \frac{1}{1-c-z-zcm_F(z)}$$

with solution

$$m_F(z) = \frac{1-c}{2cz} - \frac{1}{2c} - \frac{\sqrt{(1-c-z)^2 - 4cz}}{2cz}$$

• Using the Stieltjes inversion formula

$$f(x) \stackrel{\Delta}{=} F'(x) = \lim_{y \to 0} \frac{1}{\pi} \Im[m_F(x + iy)]$$

we finally obtain

$$f(x) = (1 - c^{-1})^+ \delta(x) + \frac{1}{2\pi cx} \sqrt{(x - a)^+ (b - x)^+}$$

with  $a = (1 - \sqrt{c})^2$ ,  $b = (1 + \sqrt{c})^2$ , of support [*a*, *b*].

J. W. Silverstein, Z. D. Bai, "On the empirical distribution of eigenvalues of a class of large dimensional random matrices," Journal of Multivariate Analysis, vol. 54, no. 2, pp. 175-192, 1995.

#### Theorem

Let  $\underline{\mathbf{B}}_N = \mathbf{X}_N \mathbf{T}_N \mathbf{X}_N^{\mathsf{H}} \in \mathbb{C}^{N \times N}$ ,  $\mathbf{X}_N \in \mathbb{C}^{N \times n}$  has i.i.d. entries of mean 0 and variance 1/N,  $F^{\mathbf{T}_N} \Rightarrow F^{\mathsf{T}}$ ,  $n/N \to c$ . Then,  $F^{\underline{\mathbf{B}}_N} \Rightarrow \underline{F}$  almost surely,  $\underline{F}$  having Stieltjes transform

$$\underline{m}_{\underline{F}}(z) = \left(c \int \frac{t}{1 + t m_{\underline{F}}(z)} dF^{T}(t) - z\right)^{-1} = \left[\frac{1}{N} \operatorname{tr} \mathbf{T}_{N} \left(\underline{m}_{\underline{F}}(z) \mathbf{T}_{N} + \mathbf{I}_{N}\right)^{-1} - z\right]^{-1}$$

which has a unique solution  $m_{\underline{F}}(z) \in \mathbb{C}^+$  if  $z \in \mathbb{C}^+$ , and  $m_{\underline{F}}(z) > 0$  if z < 0.

- in general, no explicit expression for <u>F</u>.
- Stieltjes transform of  $\mathbf{B}_N = \mathbf{T}_N^{\frac{1}{2}} \mathbf{X}_N^H \mathbf{X}_N \mathbf{T}_N^{\frac{1}{2}}$  with asymptotic distribution *F*,

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Spectrum of the sample covariance matrix model  $\mathbf{B}_N = \sum_{i=1}^n \mathbf{x}_i \mathbf{x}_i^H$ , with  $\mathbf{X}_N^H = [\mathbf{x}_1, \dots, \mathbf{x}_n]$ ,  $\mathbf{x}_i$  i.i.d. with zero mean and covariance  $\mathbf{T}_N = \mathbb{E}[\mathbf{x}_1 \mathbf{x}_1^H]$ .

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## Getting F' from $m_F$

Remember that

$$f(x) \stackrel{\Delta}{=} F'(x) = \lim_{y \to 0} \frac{1}{\pi} \Im[m_F(x + iy)]$$

• to plot the density f(x), span z = x + iy on the line  $\{x \in \mathbb{R}, y = \varepsilon\}$  parallel but close to the real axis, solve  $m_F(z)$  for each z, and plot  $\Im[m_F(z)]$ .

#### Example (Sample covariance matrix)

For *N* multiple of 3, let  $dF^T(x) = \frac{1}{3}\delta(x-1) + \frac{1}{3}\delta(x-3) + \frac{1}{3}\delta(x-K)$  and let  $\mathbf{B}_N = \mathbf{T}_N^{\frac{1}{2}}\mathbf{X}_N^H\mathbf{X}_N\mathbf{T}_N^{\frac{1}{2}}$  with  $F^{\mathbf{B}_N} \to F$ , then

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Figure: Histogram of the eigenvalues of  $\mathbf{B}_N = \mathbf{T}_N^{\frac{1}{2}} \mathbf{X}_N^H \mathbf{X}_N \mathbf{T}_N^{\frac{1}{2}}$ , N = 3000, n = 300, with  $\mathbf{T}_N$  diagonal composed of three evenly weighted masses in (i) 1, 3 and 7 on top, (ii) 1, 3 and 4 at bottom.

## The Shannon Transform

A. M. Tulino, S. Verdù, "Random matrix theory and wireless communications," Now Publishers Inc., 2004.

### Definition

Let *F* be a probability distribution,  $m_F$  its Stieltjes transform, then the Shannon-transform  $V_F$  of *F* is defined as

$$\mathcal{V}_{\mathcal{F}}(x) \stackrel{\Delta}{=} \int_{0}^{\infty} \log(1+x\lambda) dF(\lambda) = \int_{x}^{\infty} \left(\frac{1}{t} - m_{\mathcal{F}}(-t)\right) dt$$

If *F* is the distribution function of the eigenvalues of  $\mathbf{XX}^{\mathsf{H}} \in \mathbb{C}^{N \times N}$ ,

$$\mathcal{V}_F(x) = rac{1}{N} \log \det \left( \mathbf{I}_N + x \mathbf{X} \mathbf{X}^{\mathsf{H}} 
ight).$$

Note that this last relation is fundamental to wireless communication purposes!

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### Properties of the Asymptotic Support and Spiked Models

Summary of what we know and what is left to be done

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- We showed that the eigenvalue distribution  $F^{\mathbf{B}_N}$  of  $\mathbf{B}_N = \mathbf{X}\mathbf{T}\mathbf{X}^{\mathsf{H}}, F^{\mathbf{T}_N} \Rightarrow F^{\mathsf{T}}$ :
  - is similar to a deterministic  $F_N$
  - sometimes converges WEAKLY to F with Supp(F) made of compact sets.

• There is more:



For all  $N_0$ , there is no eigenvalue of  $\mathbf{B}_N$  outside  $\operatorname{Supp}(F) \cup \bigcup_{N \ge N_0} \operatorname{Supp}(F_N)$ , for all large N.

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Random Matrix Theory for Signal Processing Applicatic
## The spiked model

- For T composed of finitely many eigenvalues with large multiplicities (e.g. T = I<sub>N</sub>), no eigenvalue of B<sub>N</sub> outside Supp(F).
- If, for r fixed, T is a rank-r perturbation of I<sub>N</sub>,

diag( 
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Figure: Eigenvalues of  $\mathbf{B}_N = \mathbf{T}^{\frac{1}{2}} \mathbf{X} \mathbf{X}^{\mathsf{H}} \mathbf{T}^{\frac{1}{2}}$ , **T** diagonal of 1's but for the last four entries set to {3, 3, 2, 2}. On top, N = 500, n = 1500. At bottom, N = 500, n = 400. Theoretical limit eigenvalues of  $\mathbf{B}_N$  are stressed.

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## Limits for the spiked models

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- Assume T as above with:
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• Then, with  $\lim N/n = c$ , we have the first order limits:

$$\begin{split} \hat{\lambda}_k & \xrightarrow{\text{a.s.}} \left\{ \begin{array}{cc} 1 + \omega_k + c \frac{1 + \omega_k}{\omega_k} & , \ \omega_k > \sqrt{c} \\ (1 + \sqrt{c})^2 & , \ \omega_k \le \sqrt{c} \end{array} \right. \\ |\hat{\mathbf{u}}_k^* \mathbf{u}_k|^2 & \xrightarrow{\text{a.s.}} \left\{ \begin{array}{cc} \frac{1 - c \omega_k^{-2}}{1 + c \omega_k^{-1}} & , \ \omega_k > \sqrt{c} \\ 0 & , \ \omega_k \le \sqrt{c} \end{array} \right. \end{split}$$

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• As well as the second order limits in the Gaussian case:

$$\sqrt{N} \begin{pmatrix} |\mathbf{u}_{k}^{*}\hat{\mathbf{u}}_{k}|^{2} - \left[\frac{1-c\omega_{k}^{-2}}{1+c\omega_{k}^{-1}}\right] \\ \hat{\lambda}_{k} - \left[1+\omega_{k}+c\frac{1+\omega_{k}}{\omega_{k}}\right] \end{pmatrix} \Rightarrow \mathcal{CN} \begin{pmatrix} 0, \left[\frac{c^{2}(1+\omega_{k})^{2}}{(c+\omega_{k})^{2}(\omega_{k}^{2}-c)}\left(\frac{c(1+\omega_{k})^{2}}{(c+\omega_{k})^{2}\omega_{k}^{2}}+1\right) & \frac{(1+\omega_{k})^{3}c^{2}}{(\omega_{k}+c)^{2}\omega_{k}} \\ \frac{(1+\omega_{k})^{3}c^{2}}{(\omega_{k}+c)^{2}\omega_{k}} & \frac{c(1+\omega_{k})^{2}(\omega_{k}^{2}-c)}{\omega_{k}^{2}} \\ \end{pmatrix} \end{pmatrix}$$
If  $\omega_{k} < \sqrt{c}$ 

$$N^{\frac{2}{3}} \frac{\lambda_k - (1 + \sqrt{c})^2}{(1 + \sqrt{c})^{\frac{4}{3}} \sqrt{c}} \Rightarrow T_2$$

with  $T_2$  the complex Tracy-Widom distribution.

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# Second order statistics, $\omega_k < \sqrt{c}$



Figure: Distribution of  $N^{\frac{2}{3}}c^{-\frac{1}{2}}(1+\sqrt{c})^{-\frac{4}{3}}\left[\hat{\lambda}_{k}-(1+\sqrt{c})^{2}\right]$  against the Tracy-Widom law for N = 500, n = 1500, c = 1/3,  $\mathbf{T} = \text{diag}(1, \dots, 1, 1.5)$  ( $0.5 < \sqrt{c}$ ). Empirical distribution taken over 10,000 Monte-Carlo simulations.

# Second order statistics, $\omega_k > \sqrt{c}$



Figure: Empirical and theoretical distribution of the fluctuations of  $\hat{\mathbf{u}}_1$  if *X* has i.i.d.  $\mathcal{CN}(0, 1/n)$  entries, N/n = 1/8, N = 64,  $\omega_1 = 1$  (left) or  $\omega_1 = 0.5$  (right).

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  - sample covariance matrix models,  $\mathbf{XTX}^{H}$  and  $\mathbf{T}^{\frac{1}{2}}\mathbf{X}^{H}\mathbf{XT}^{\frac{1}{2}}$
  - doubly correlated models, R<sup>1</sup>/<sub>2</sub> XTX<sup>H</sup>R<sup>1</sup>/<sub>2</sub>. With X Gaussian, Kronecker model.
  - doubly correlated models with external matrix,  $\mathbf{R}^{\frac{1}{2}} \mathbf{X} \mathbf{T} \mathbf{X}^{\mathsf{H}} \mathbf{R}^{\frac{1}{2}} + \mathbf{A}$ .
  - variance profile, **XX**<sup>H</sup>, where **X** has i.i.d. entries with mean 0, variance  $\sigma_{i,i}^2$
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  - sum of doubly correlated i.i.d. matrices,  $\sum_{k=1}^{K} \mathbf{R}_{k}^{\frac{1}{2}} \mathbf{X}_{k} \mathbf{T}_{k} \mathbf{X}_{k}^{H} \mathbf{R}_{k}^{\frac{1}{2}}$ .
  - information-plus-noise models (X + A)(X + A)<sup>H</sup>
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- R- and S-transforms: models involving a column subset W of unitary matrices
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In most cases, **T** and **R** can be taken random, but independent of **X**. More involved random matrices, such as Vandermonde matrices, were not yet studied.

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# Models studied with analytic tools

- Stieltjes transform: models involving i.i.d. matrices
  - sample covariance matrix models,  $\mathbf{XTX}^{H}$  and  $\mathbf{T}^{\frac{1}{2}}\mathbf{X}^{H}\mathbf{XT}^{\frac{1}{2}}$
  - doubly correlated models, R<sup>1</sup>/<sub>2</sub> XTX<sup>H</sup>R<sup>1</sup>/<sub>2</sub>. With X Gaussian, Kronecker model.
  - doubly correlated models with external matrix,  $\mathbf{R}^{\frac{1}{2}} \mathbf{X} \mathbf{T} \mathbf{X}^{\mathsf{H}} \mathbf{R}^{\frac{1}{2}} + \mathbf{A}$ .
  - variance profile, **XX**<sup>H</sup>, where **X** has i.i.d. entries with mean 0, variance  $\sigma_{i,i}^2$
  - Ricean channels,  $\mathbf{X}\mathbf{X}^{H} + \mathbf{A}$ , where **X** has a variance profile.
  - sum of doubly correlated i.i.d. matrices,  $\sum_{k=1}^{K} \mathbf{R}_{k}^{\frac{1}{2}} \mathbf{X}_{k} \mathbf{T}_{k} \mathbf{X}_{k}^{H} \mathbf{R}_{k}^{\frac{1}{2}}$ .
  - information-plus-noise models (X + A)(X + A)<sup>H</sup>
  - frequency-selective doubly-correlated channels  $(\sum_{k=1}^{K} \mathbf{R}_{k}^{\frac{1}{2}} \mathbf{X}_{k} \mathbf{T}_{k} \mathbf{X}_{k} \mathbf{R}_{k}^{\frac{1}{2}}) (\sum_{k=1}^{K} \mathbf{R}_{k}^{\frac{1}{2}} \mathbf{X}_{k} \mathbf{T}_{k} \mathbf{X}_{k} \mathbf{R}_{k}^{\frac{1}{2}})$
  - sum of frequency-selective doubly-correlated channels  $\sum_{k=1}^{K} \mathbf{R}_{k}^{\frac{1}{2}} \mathbf{H}_{k} \mathbf{T}_{k} \mathbf{H}_{k}^{H} \mathbf{R}_{k}^{\frac{1}{2}}$ , where  $\mathbf{H}_{k} = \sum_{l=1}^{L} \mathbf{R}_{kl}^{\prime \frac{1}{2}} \mathbf{X}_{kl} \mathbf{T}_{kl}^{\prime} \mathbf{X}_{kl}^{H} \mathbf{R}_{kl}^{\prime \frac{1}{2}}$ .
- R- and S-transforms: models involving a column subset W of unitary matrices
  - doubly correlated Haar matrix  $\mathbf{R}^{\frac{1}{2}} \mathbf{W} \mathbf{T} \mathbf{W}^{\mathsf{H}} \mathbf{R}^{\frac{1}{2}}$
  - sum of simply correlated Haar matrices  $\sum_{k=1}^{K} \mathbf{W}_k \mathbf{T}_k \mathbf{W}_k^H$

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# Tools for Random Matrix Theory Summary of what we know and what is left to be done Models studied with moments/free probability

### asymptotic results

- most of the above models with Gaussian X.
- products V<sub>1</sub>V<sub>1</sub><sup>H</sup>T<sub>1</sub>V<sub>2</sub>V<sub>2</sub><sup>H</sup>T<sub>2</sub>... of Vandermonde and deterministic matrices
- conjecture: any probability space of matrices invariant to row or column permutations.
- marginal studies, not yet fully explored
  - rectangular free convolution: singular values of rectangular matrices
  - finite size models. Instead of almost sure convergence of  $m_{X_N}$  as  $N \to \infty$ , we can study finite size behaviour of  $E[m_{X_N}]$ .

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- Classical Random Matrix Theory
- Introduction to Large Dimensional Random Matrix Theory
- The Random Matrix Pioneers
- The Moment Approach and Free Probability
- Introduction of the Stieltjes Transform
- Properties of the Asymptotic Support and Spiked Models
- Summary of what we know and what is left to be done

### 2 Random Matrix Theory and Signal Source Sensing

- Small Dimensional Analysis
- Large Dimensional Random Matrix Analysis

### 3 Random Matrix Theory and Multi-Source Power Estimation

- Optimal detector
- The moment method
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### 4 Random Matrix Theory and Failure Detection in Complex Systems

- Random matrix models of local failures in sensor networks
- Failure detection and localization

# Random Matrix Theory and Signal Source Sensing Signal Sensing in Cognitive Radios



Secondary Network

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### Assume the scenario of

- an *hypothetical* signal source  $\sqrt{P}\mathbf{x} \in \mathbb{C}^n$  of power *P*
- a transfer channel  $\mathbf{H} \in \mathbb{C}^{N \times n}$
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- additive noise  $\sigma \mathbf{w} \in \mathbb{C}^N$  of variance  $\sigma^2 \mathbf{I}_N$ .
- We consider the following hypothesis test

$$\mathbf{y}^{(m)} = \begin{cases} \sigma \mathbf{w}^{(m)} & , (\mathcal{H}_0) \\ \sqrt{P} \mathbf{H} \mathbf{x}^{(m)} + \sigma \mathbf{w}^{(m)} & , (\mathcal{H}_1) \end{cases}$$

- We wish to confront the hypotheses  $\mathcal{H}_0$  and  $\mathcal{H}_1$  given the data matrix  $\mathbf{Y} \triangleq [\mathbf{y}^{(1)}, \dots, \mathbf{y}^{(M)}] \in \mathbb{C}^{N \times M}$ .
- We consider, in a Bayesian framework, the Neyman-Pearson test ratio

$$C(\mathbf{Y}) \triangleq \frac{P_{\mathcal{H}_1|\mathbf{Y},l}(\mathbf{Y})}{P_{\mathcal{H}_0|\mathbf{Y},l}(\mathbf{Y})}$$

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with prior information *I* on **H**,  $\mathbf{x}^{(m)}$ ,  $\sigma$ , ....

### We assume prior statistical and deterministic knowledge I on H, σ, P

• Using the maximum entropy principle (MaxEnt), a prior  $P_{(\mathbf{H},\sigma,P)}(\mathbf{H},\sigma,P)$  can be derived

$$P_{\mathbf{Y}|\mathcal{H}_{j},l}(\mathbf{Y}) = \int_{(\mathbf{H},\sigma,P)} P_{\mathbf{Y}|\mathcal{H}_{j},l,\mathbf{H},\sigma,P}(\mathbf{Y}) P_{(\mathbf{H},\sigma,P)}(\mathbf{H},\sigma,P) d(\mathbf{H},\sigma,P)$$

- In the following.
  - we derive the case P = 1,  $\sigma$  known and the knowledge about **H** conveys unitary invariance
    - E[tr HH<sup>H</sup>] known: this is what we assume here;
    - E[HH<sup>H</sup>] = Q unknown but such that E[tr Q] is known;
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# Evaluation of $P_{\mathbf{Y}|\mathcal{H}_i,l}(\mathbf{Y})$

- by MaxEnt, **X**, **W** are standard Gaussian matrix with  $X_{ij}$ ,  $W_{ij} \sim CN(0, 1)$ .
- Under *H*<sub>0</sub>:
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with  $\Sigma = E[\mathbf{y}^{(1)}\mathbf{y}^{(1)H}] = \mathbf{H}\mathbf{H}^{H} + \sigma^{2}\mathbf{I}_{N}$ . From unitary invariance of **H**, denoting  $\Sigma = \mathbf{U}\mathbf{G}\mathbf{U}^{H}$ , diag(**G**) =  $(g_{1}, \dots, g_{n}, \sigma^{2}, \dots, \sigma^{2})$ 

$$P_{\mathbf{Y}|\mathcal{H}_{1}}(\mathbf{Y}) = \int_{\mathcal{U}(N) \times (\sigma^{2}, \infty)^{n}} P_{\mathbf{Y}|\mathbf{U}\mathbf{G}\mathbf{U}^{\mathsf{H}}, \mathcal{H}_{1}}(\mathbf{Y}, \mathbf{U}, \mathbf{G}) P_{\mathbf{U}}(\mathbf{U}) P_{(g_{1}, \dots, g_{n})}(g_{1}, \dots, g_{n}) d\mathbf{U} dg_{1} \dots dg_{n}$$

where

- $P_{\mathbf{Y}|\mathbf{U}\mathbf{G}\mathbf{U}^{H},\mathcal{H}_{1}}$  is Gaussian with zero mean and variance  $\mathbf{U}\mathbf{G}\mathbf{U}^{H}$ ;
- P<sub>U</sub> is a constant (d**U** is a Haar measure);
- if **H** is Gaussian,  $P_{(a_1-\sigma^2,\ldots,a_n-\sigma^2)}$  is the joint eigenvalue distribution of a central Wishart;

Evaluation of  $P_{\mathbf{Y}|\mathcal{H}_{i},l}(\mathbf{Y})$ 

- by MaxEnt, **X**, **W** are standard Gaussian matrix with  $X_{ij}$ ,  $W_{ij} \sim CN(0, 1)$ .
- **Under** *H*<sub>0</sub>:
  - $\mathbf{Y} = \sigma \mathbf{W}$

$$\mathcal{P}_{\mathbf{Y}|\mathcal{H}_0,I}(\mathbf{Y}) = rac{1}{(\pi\sigma^2)^{NM}} e^{-rac{1}{\sigma^2}\operatorname{tr}\mathbf{Y}\mathbf{Y}^{\mathsf{H}}}.$$

• **Under** *H*<sub>1</sub>:

• 
$$\mathbf{Y} = \begin{bmatrix} \sqrt{P}\mathbf{H} & \sigma \mathbf{I}_N \end{bmatrix} \begin{bmatrix} \mathbf{X} \\ \mathbf{W} \end{bmatrix}$$
  
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R. Couillet, M. Debbah, "A Bayesian Framework for Collaborative Multi-Source Signal Sensing", IEEE Transactions on Signal Processing, vol. 58, no. 10, pp. 5186-5195, 2010.

### Theorem (Neyman-Pearson test)

The ratio  $C(\mathbf{Y})$  when the receiver knows n = 1, P = 1,  $E[\frac{1}{N} \text{ tr } \mathbf{H}\mathbf{H}^{H}] = 1$  and  $\sigma^{2}$ , reads

$$C(\mathbf{Y}) = \frac{1}{N} \sum_{l=1}^{N} \frac{\sigma^{2(N+M-1)} e^{\sigma^2 + \frac{\lambda_l}{\sigma^2}}}{\prod_{\substack{i=1\\i\neq l}}^{N} (\lambda_l - \lambda_i)} J_{N-M-1}(\sigma^2, \lambda_l)$$

with  $\lambda_1, \ldots, \lambda_N$  the eigenvalues of **YY**<sup>H</sup> and where

$$J_k(x,y) \triangleq \int_x^{+\infty} t^k e^{-t - \frac{y}{t}} dt.$$

- non trivial dependency on  $\lambda_1, \ldots, \lambda_N$
- contrary to energy detector,  $\sum_i \lambda_i$  is not a sufficient statistic;
- integration over  $\sigma^2$  (or *P* when  $P \neq 1$ ) is difficult.

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Comparison to energy detector



Figure: ROC curve for single-source detection, K = 1, N = 4, M = 8, SNR = -3 dB, FAR range of practical interest, with signal power P = 0 dBm, either known or unknown at the receiver.
# Comparison to energy detector



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# Large Dimensional Random Matrix Analysis

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If  $\mathcal{H}_0$ , then the eigenvalues of  $\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\mathsf{H}} = \sigma^2 \frac{1}{N}\mathbf{W}\mathbf{W}^{\mathsf{H}}$  asymptotically distribute as



Figure: Marčenko-Pastur law with  $c = \lim N/L$ .

## Reminder:

#### Theorem

*P*(no eigenvalues outside  $[\sigma^2(1-\sqrt{c})^2, \sigma^2(1+\sqrt{c})^2]$  for all large *N*) = 1

• If  $\mathcal{H}_0$ ,

$$\frac{\lambda_{\max}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\mathsf{H}})}{\lambda_{\min}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\mathsf{H}})} \xrightarrow{\text{a.s.}} \frac{(1+\sqrt{c})^2}{(1-\sqrt{c})^2}$$

independent of the SNR!

Sac

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Conditioning number test

$$C_{\text{cond}}(\mathbf{Y}) = \frac{\lambda_{\max}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\mathsf{H}})}{\lambda_{\min}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\mathsf{H}})}$$

- if  $C_{\text{cond}}(\mathbf{Y}) > \tau$ , presence of a signal.
- if  $C_{\text{cond}}(\mathbf{Y}) < \tau$ , absence of signal.
- but this is *ad-hoc*! how good does it compare to optimal?
- can we find non *ad-hoc* approaches?

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#### Generalized Likelihood Ratio Test

Alternative test to Neyman-Pearson test,

$$C_{\text{GLRT}}(\mathbf{Y}) = \frac{\sup_{\mathbf{H},\sigma^2} P_{\mathcal{H}_1|\mathbf{Y},\mathbf{H},\sigma^2}(\mathbf{Y})}{\sup_{\sigma^2} P_{\mathcal{H}_0|\mathbf{Y},\sigma^2}(\mathbf{Y})}$$

- based on ratios of maximum likelihood
- clearly sub-optimal but avoid the need for priors.
- GLRT test

$$C_{\text{GLRT}}(\mathbf{Y}) = \left( \left(1 - \frac{1}{N}\right)^{N-1} \frac{\lambda_{\max}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\text{H}})}{\frac{1}{N}\sum_{i=1}^{N}\lambda_i} \left(1 - \frac{\lambda_{\max}(\frac{1}{N}\mathbf{Y}\mathbf{Y}^{\text{H}})}{\sum_{i=1}^{N}\lambda_i}\right)^{N-1} \right)^{-1}$$

Contrary to the ad-hoc conditioning number test, GLRT based on



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Figure: ROC curve for *a priori* unknown  $\sigma^2$  of the Bayesian method, conditioning number method and GLRT method, M = 1, N = 4, L = 8, SNR = 0 dB. For the Bayesian method, both uniform and Jeffreys prior, with exponent  $\alpha = 1$ , are provided.

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## Random Matrix Theory and Multi-Source Power Estimation Application Context: Coverage range in Femtocells



Secondary Network

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$$\mathbf{y}^{(m)} = \sum_{k=1}^{K} \sqrt{P_k} \mathbf{H}_k \mathbf{x}_k^{(m)} + \sigma \mathbf{w}^{(m)}$$

and wish to infer  $P_1, \ldots, P_K$ .

• With  $\mathbf{Y} = [\mathbf{y}^{(1)}, \dots, \mathbf{y}^{(M)}]$ , this can be rewritten

$$\mathbf{Y} = \sum_{k=1}^{K} \sqrt{P_k} \mathbf{H}_k \mathbf{X}_k + \sigma \mathbf{W} = \underbrace{\left[\sqrt{P_1} \mathbf{H}_1 \cdots \sqrt{P_K} \mathbf{H}_K\right]}_{\triangleq \mathbf{HP}^{\frac{1}{2}}} \underbrace{\begin{bmatrix} \mathbf{X}_1 \\ \vdots \\ \mathbf{X}_K \end{bmatrix}}_{\triangleq \mathbf{X}} + \sigma \mathbf{W} = \begin{bmatrix} \mathbf{HP}^{\frac{1}{2}} & \sigma \mathbf{I}_N \end{bmatrix} \begin{bmatrix} \mathbf{X} \\ \mathbf{W} \end{bmatrix}.$$

• If H, (X<sup>T</sup> W<sup>T</sup>) are unitarily invariant, Y is unitarily invariant.

Most information about  $P_1, \ldots, P_K$  is contained in the eigenvalues of  $\mathbf{B}_N \triangleq \frac{1}{M} \mathbf{Y} \mathbf{Y}^H$ .

Sac

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#### Random Matrix Theory and Multi-Source Power Estimation From small to large system analysis



The classical approach requires to evaluate  $P_{P_1,...,P_K|Y}$ 

- assuming Gaussian parameters, this is similar to previous calculus
- leads to a very involved expression
- prohibitively expensive to evaluate even for small N, n<sub>k</sub>, M

#### Random Matrix Theory and Multi-Source Power Estimation From small to large system analysis



Assuming dimensions N,  $n_k$ , M grow large, large dimensional random matrix theory provides

- a link between:
  - the "observation": the limiting spectral distribution (l.s.d.) of **B**<sub>N</sub>;
  - the "hidden parameters": the powers  $P_1, \ldots, P_K$ , i.e. the l.s.d. of **P**.
- consistent estimators of the hidden parameters.

Image: A matrix

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R. Couillet and M. Guillaud, "Performance of Statistical Inference Methods for the Energy Estimation of Multiple Sources," *Invited Paper*, IEEE International Communications Conference, Nice, France, 2011.

conditional probability

#### Theorem

Assume  $P_1, \ldots, P_K$  have multiplicity  $n_1 = \ldots = n_K = 1$ . Then, denoting  $\lambda = (\lambda_1, \ldots, \lambda_N)$  the eigenvalues of  $B_N$ 

$$P_{\mathbf{Y}|P_{1},...,P_{K}}(\mathbf{Y}) = \frac{C(-1)^{Nn+1}e^{N\sigma^{2}\sum_{i=1}^{n}\frac{P_{i}}{P_{i}}}}{\sigma^{2(N-n)(M-n)}\prod_{i=1}^{n}P_{i}^{M-n+1}\Delta(\mathbf{P})}\sum_{\mathbf{a}\in\mathcal{S}_{n}^{N}}(-1)^{|\mathbf{a}|}\operatorname{sgn}(\mathbf{a})e^{\frac{M}{\sigma^{2}}|\lambda[\tilde{\mathbf{a}}]|} \\ \times \frac{\Delta(\operatorname{diag}(\boldsymbol{\lambda}[\tilde{\mathbf{a}}]))}{\Delta(\operatorname{diag}(\boldsymbol{\lambda}))}\sum_{\mathbf{b}\in\mathcal{S}_{n}}\operatorname{sgn}(\mathbf{b})\prod_{i=1}^{n}J_{N-M-1}\left(\frac{N\sigma^{2}}{P_{b_{i}}},\frac{NM\lambda_{a_{i}}}{P_{b_{i}}}\right).$$

ML/MMSE estimators

$$\underline{\hat{P}}^{(\mathrm{ML})} = \arg \max_{P_1, \dots, P_K} P_{\mathbf{Y}|P_1, \dots, P_K}(\mathbf{Y})$$

$$\underline{\hat{P}}^{(\mathrm{MMSE})} = \int_{[0,\infty)^K} (P_1, \dots, P_K) P_{P_1, \dots, P_K|\mathbf{Y}}(P_1, \dots, P_K) dP_1 \dots dP_K$$

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Random Matrix Theory for Signal Processing Application

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Assume  $P_1, \ldots, P_K$  have multiplicity  $n_1 = \ldots = n_K = 1$ . Then, denoting  $\lambda = (\lambda_1, \ldots, \lambda_N)$  the eigenvalues of  $B_N$ 

$$P_{\mathbf{Y}|P_{1},...,P_{K}}(\mathbf{Y}) = \frac{C(-1)^{Nn+1}e^{N\sigma^{2}\sum_{i=1}^{n}\frac{P_{i}}{P_{i}}}}{\sigma^{2(N-n)(M-n)}\prod_{i=1}^{n}P_{i}^{M-n+1}\Delta(\mathbf{P})}\sum_{\mathbf{a}\in\mathcal{S}_{n}^{N}}(-1)^{|\mathbf{a}|}\operatorname{sgn}(\mathbf{a})e^{\frac{M}{\sigma^{2}}|\lambda[\mathbf{\tilde{a}}]|} \\ \times \frac{\Delta(\operatorname{diag}(\lambda[\mathbf{\tilde{a}}]))}{\Delta(\operatorname{diag}(\lambda))}\sum_{\mathbf{b}\in\mathcal{S}_{n}}\operatorname{sgn}(\mathbf{b})\prod_{i=1}^{n}J_{N-M-1}\left(\frac{N\sigma^{2}}{P_{b_{i}}},\frac{NM\lambda_{a_{i}}}{P_{b_{i}}}\right).$$

ML/MMSE estimators

$$\underline{\hat{P}}^{(\mathrm{ML})} = \arg \max_{P_1, \dots, P_K} P_{\mathbf{Y}|P_1, \dots, P_K}(\mathbf{Y})$$

$$\underline{\hat{P}}^{(\mathrm{MMSE})} = \int_{[0,\infty)^K} (P_1, \dots, P_K) P_{P_1, \dots, P_K} |\mathbf{Y}(P_1, \dots, P_K) dP_1 \dots dP_K$$

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- Introduction to Large Dimensional Random Matrix Theory
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- Introduction of the Stieltjes Transform
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- Summary of what we know and what is left to be done

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- Large Dimensional Random Matrix Analysis

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- Failure detection and localization

(I)

Reminder on free deconvolution

Free probability provides tools to compute

$$p_k = \frac{1}{K} \sum_{i=1}^{K} \lambda(\mathbf{P})^k = \frac{1}{K} \sum_{i=1}^{K} P_i^k$$

as a function of

$$b_k = rac{1}{N}\sum_{i=1}^N \lambda(rac{1}{M}\mathbf{Y}\mathbf{Y}^\mathsf{H})^k$$

- One can obtain all the successive sum powers of P<sub>1</sub>,..., P<sub>K</sub>.
- From that, we can infer on the values of each P<sub>k</sub>
- The tools come from the relations,
  - cumulant to moment (and also moment to cumulant)

$$M_n = \sum_{\pi \in NC(n)} \prod_{V \in \pi} C_{|V|}$$

• Sums of cumulants for asymptotically free A and B (of measure  $\mu_A \boxplus \mu_B$ ),

$$C_k(\mathbf{A} + \mathbf{B}) = C_k(\mathbf{A}) + C_k(\mathbf{B})$$

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• Moments of information plus noise models  $\mathbf{B}_N = \frac{1}{n} (\mathbf{A}_N + \sigma \mathbf{W}_N) (\mathbf{A}_N + \sigma \mathbf{W}_N)^H$ ,

$$\mu_{B} = \left( \left( \mu_{A} \boxtimes \mu_{c} \right) \boxplus \delta_{\sigma^{2}} \right) \boxtimes \mu_{c}$$
with ratio c

with  $\mu_c$  the Marčenko-Pastur law with ratio c

R. Couillet (Supélec)

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R. Couillet (Supélec

Random Matrix Theory for Signal Processing Application

# • one can deconvolve **YY**<sup>H</sup> in three steps,

• an information-plus-noise model with "deterministic matrix"  $HP^{\frac{1}{2}}XX^{H}P^{\frac{1}{2}}H^{H}$ ,

$$\mathbf{Y}\mathbf{Y}^{\mathsf{H}} = (\mathbf{H}\mathbf{P}^{\frac{1}{2}}\mathbf{X} + \sigma\mathbf{W})(\mathbf{H}\mathbf{P}^{\frac{1}{2}}\mathbf{X} + \sigma\mathbf{W})^{\mathsf{H}}$$

- from  $\mathbf{HP}^{\frac{1}{2}}\mathbf{XX}^{H}\mathbf{P}^{\frac{1}{2}}\mathbf{H}^{H}$ , up to a Gram matrix commutation, we can deconvolve the signal **X**,  $\mathbf{P}^{\frac{1}{2}}\mathbf{HH}^{H}\mathbf{P}^{\frac{1}{2}}\mathbf{XX}^{H}$
- from  $\mathbf{P}^{\frac{1}{2}}\mathbf{H}\mathbf{H}^{H}\mathbf{P}^{\frac{1}{2}}$ , a new matrix commutation allows one to deconvolve  $\mathbf{H}\mathbf{H}^{H}$

PHH<sup>H</sup>

Sac

- one can deconvolve **YY**<sup>H</sup> in three steps,
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- from P<sup>1/2</sup> HH<sup>H</sup>P<sup>1/2</sup>, a new matrix commutation allows one to deconvolve HH<sup>H</sup>
   PHH<sup>H</sup>

500

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500

In terms of distributions

$$\begin{split} \mu^{\infty}_{\frac{1}{M}\mathsf{HP}^{\frac{1}{2}}\mathsf{X}\mathsf{X}^{\mathsf{HP}^{\frac{1}{2}}}\mathsf{H}^{\mathsf{H}}} &= \left( \left( \mu^{\infty}_{\mathsf{B}_{\mathsf{N}}} \boxtimes \mu_{\frac{1}{c}} \right) \boxminus \delta_{\sigma^{2}} \right) \boxtimes \mu_{\frac{1}{c}} \\ \mu^{\infty}_{\mathsf{P}^{\frac{1}{2}}}\mathsf{H}^{\mathsf{H}}\mathsf{HP}^{\frac{1}{2}} &= \mu^{\infty}_{\frac{1}{M}\mathsf{P}^{\frac{1}{2}}}\mathsf{H}^{\mathsf{H}}\mathsf{HP}^{\frac{1}{2}}\mathsf{X}\mathsf{X}^{\mathsf{H}}} \boxtimes \mu_{\frac{1}{c_{0}}} \\ \mu^{\infty}_{\mathsf{P}} &= \mu^{\infty}_{\mathsf{P}\mathsf{H}^{\mathsf{H}}\mathsf{H}} \boxtimes \mu_{\frac{1}{c_{0}}} \end{split}$$

• Numerically, with  $b_m \triangleq \frac{1}{N} \mathbb{E} \left[ \text{tr } \mathbf{B}_N^m \right]$  and  $p_m \triangleq \sum_{k=1}^K \frac{n_k}{n} P_k^m$ 

$$\begin{split} & b_{1} = N^{-1}np_{1} + 1 \\ & b_{2} = \left(N^{-2}M^{-1}n + N^{-1}n\right)p_{2} + \left(N^{-2}n^{2} + N^{-1}M^{-1}n^{2}\right)p_{1}^{-2} + \left(2N^{-1}n + 2M^{-1}n\right)p_{1} + \left(1 + NM^{-1}\right) \\ & b_{3} = \left(3N^{-3}M^{-2}n + N^{-3}n + 6N^{-2}M^{-1}n + N^{-1}M^{-2}n + N^{-1}n\right)p_{3} \\ & + \left(6N^{-3}M^{-1}n^{2} + 6N^{-2}M^{-2}n^{2} + 3N^{-2}n^{2} + 3N^{-1}M^{-1}n^{2}\right)p_{2}p_{1} \\ & + \left(N^{-3}M^{-2}n^{3} + N^{-3}n^{3} + 3N^{-2}M^{-1}n^{3} + N^{-1}M^{-2}n^{3}\right)p_{1}^{3} \\ & + \left(6N^{-2}M^{-1}n + 6N^{-1}M^{-2}n + 3N^{-1}n + 3M^{-1}n\right)p_{2} \\ & + \left(3N^{-2}M^{-2}n^{2} + 3N^{-2}n^{2} + 9N^{-1}M^{-1}n^{2} + 3M^{-2}n^{2}\right)p_{1}^{2} \\ & + \left(3N^{-1}M^{-2}n + 3N^{-1}n + 9M^{-1}n + 3NM^{-2}n\right)p_{1}. \end{split}$$

R. Couillet (Supélec)

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• Once the  $p_i^m$  are obtained, in the particular case  $n_1 = \ldots = n_K$ , Newton-Girard formulas give  $P_1, \ldots, P_K$  as the solutions of

$$X^{K} - \Pi_{1} X^{K-1} + \Pi_{2} X^{K-2} - \ldots + (-1)^{K} \Pi_{K} = 0$$

with  $\Pi_1, \ldots, \Pi_n$  recursively computed from

$$(-1)^{K} K \Pi_{K} + \sum_{i=1}^{K} (-1)^{K+i} p_{i} \Pi_{K-i} = 0.$$

- fast method but with major limitations!
  - polynomial solutions can be purely complex
  - moment estimates propagate errors to higher order moments (2nd estimate 10<sup>3</sup> worse than 1st!)
  - modifying Newton-Girard formulas boils down to ad-hoc methods..
  - ML and MMSE methods are prohibitively expensive.

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Recall the model

$$\mathbf{Y} = \begin{bmatrix} \mathbf{HP}^{\frac{1}{2}} & \sigma \mathbf{I}_N \end{bmatrix} \begin{bmatrix} \mathbf{X} \\ \mathbf{W} \end{bmatrix}$$

#### very similar to a sample covariance matrix.

for simplicity of analysis, consider the sample covariance matrix model

$$\mathbf{Y}_{=}^{\Delta} \mathbf{T}^{\frac{1}{2}} \mathbf{X} \in \mathbb{C}^{N \times n}, \ \mathbf{B}_{N} = \frac{1}{n} \mathbf{Y} \mathbf{Y}^{\mathsf{H}} \in \mathbb{C}^{N \times N}, \ \underline{\mathbf{B}}_{N} = \frac{1}{n} \mathbf{Y}^{\mathsf{H}} \mathbf{Y} \in \mathbb{C}^{n \times n}$$

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• If  $F^{\mathsf{T}} \Rightarrow T$ , then  $m_{F^{\mathsf{B}}N}(z) = m_{\mathsf{B}}(z) \xrightarrow{\text{a.s.}} m_{F}(z)$  such that

$$m_{\underline{F}}(z) = \left(c \int \frac{t}{1 + tm_{\underline{F}}(z)} dT(t) - z\right)^{-1}$$
  
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Sac

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$$\mathbf{t}_{\mathbf{k}} = \frac{1}{2\pi i} \oint_{\mathcal{C}_{\mathbf{k}}} \frac{\omega}{\mathbf{t}_{\mathbf{k}} - \omega} d\omega = \frac{1}{2\pi i} \oint_{\mathcal{C}_{\mathbf{k}}} \frac{1}{N_{\mathbf{k}}} \sum_{j=1}^{K} N_{j} \frac{\omega}{\mathbf{t}_{j} - \omega} d\omega = \frac{N}{2\pi i N_{\mathbf{k}}} \oint_{\mathcal{C}_{\mathbf{k}}} \omega m_{T}(\omega) d\omega.$$

• After the variable change  $\omega = -1/m_{\underline{F}}(z)$ ,

$$t_{k} = \frac{N}{N_{k}} \frac{1}{2\pi i} \oint_{C_{E,k}} zm_{F}(z) \frac{m'_{E}(z)}{m^{2}_{F}(z)} dz$$

• When the system dimensions are large,

$$m_{\mathsf{F}}(z) \simeq m_{\mathsf{B}_N}(z) \stackrel{\Delta}{=} \frac{1}{N} \sum_{k=1}^N \frac{1}{\lambda_k - z}, \quad \text{with} \quad (\lambda_1, \dots, \lambda_N) = \operatorname{eig}(\mathsf{B}_N) = \operatorname{eig}(\frac{1}{n} \mathsf{Y} \mathsf{Y}^{\mathsf{H}}).$$

Dominated convergence arguments then show

$$t_k - \hat{t}_k \xrightarrow{\text{a.s.}} 0 \text{ with } \hat{t}_k = \frac{N}{N_k} \frac{1}{2\pi i} \oint_{\mathcal{C}_{\underline{E},k}} z m_{\mathbf{B}_N}(z) \frac{m_{\underline{B}_N}'(z)}{m_{\underline{B}_N}^2(z)} dz.$$

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$$t_{k} = \frac{1}{2\pi i} \oint_{\mathcal{C}_{k}} \frac{\omega}{t_{k} - \omega} d\omega = \frac{1}{2\pi i} \oint_{\mathcal{C}_{k}} \frac{1}{N_{k}} \sum_{j=1}^{K} N_{j} \frac{\omega}{t_{j} - \omega} d\omega = \frac{N}{2\pi i N_{k}} \oint_{\mathcal{C}_{k}} \omega m_{T}(\omega) d\omega.$$

• After the variable change  $\omega = -1/m_{\underline{F}}(z)$ ,

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When the system dimensions are large,

$$m_F(z) \simeq m_{\mathbf{B}_N}(z) \stackrel{\Delta}{=} \frac{1}{N} \sum_{k=1}^N \frac{1}{\lambda_k - z}, \quad \text{with} \quad (\lambda_1, \dots, \lambda_N) = \operatorname{eig}(\mathbf{B}_N) = \operatorname{eig}(\frac{1}{n} \mathbf{Y} \mathbf{Y}^{\mathsf{H}}).$$

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## Intuition:

- m<sub>E</sub>(z) is defined outside the support of <u>E</u>
- on the real axis,  $m'_{\underline{F}}(z) = \int \frac{1}{(t-z)^2} d\underline{F}(t) > 0$
- it therefore has a local growing inverse outside the support of <u>F</u>
- notice that  $m_F(z)$  has a closed-form inverse

$$z_{\underline{F}}(m) = -\frac{1}{m} + c \int \frac{t}{1+tm} dT(t)$$

It can be shown that  $z_F(m)$ , m < 0, is growing *if and only if* its image is outside the support of <u>F</u>.

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Random Matrix Theory and Multi-Source Power Estimation The Stielties transform metho Inverse formula for the Stieltjes transform



Figure:  $z_E(m)$ , with <u>*F*</u> the l.s.d. of <u>**B**</u><sub>N</sub> = **X**<sub>N</sub><sup>H</sup>**T**<sub>N</sub>**X**<sub>N</sub> with **T**<sub>N</sub> diagonal composed of three evenly weighted masses in 1, 3 and 7. The support of *F* is read on the vertical axis, whenever  $x_E(m)$  is not increasing.

• denote  $x_k^-$ ,  $x_k^+$  two points on either side of cluster k in  $\underline{F}$  such that  $x_k^- = \underline{z}_{\underline{F}}(m_k^-)$  and  $x_k^+ = \underline{z}_{\underline{F}}(m_k^+)$ .

from the asymptotes, we observe that

$$t_{k-1} < -\frac{1}{m_k^-} < t_k < -\frac{1}{m_k^+} < t_{k+1}$$

• we can therefore take a contour  $C_{\underline{E},k}$  that crosses the real line at  $-\frac{1}{m_k^-}$  and at  $-\frac{1}{m_k^+}$  and is outside the real line everywhere else.

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X. Mestre, "Improved estimation of eigenvalues and eigenvectors of covariance matrices using their sample estimates," IEEE trans. on Information Theory, vol. 54, no. 11, pp. 5113-5129, 2008.

• If remains to compute the integral from residue calculus.

$$\hat{\mathbf{t}}_{k} = \frac{N}{N_{k}} \frac{1}{2\pi i} \oint_{\mathcal{C}_{\underline{E},k}} z m_{\mathbf{B}_{N}}(z) \frac{m'_{\underline{B}_{N}}(z)}{m'_{\underline{B}_{N}}(z)} dz.$$

- From exact separation (Bai and Silverstein, 1998), C<sub>E,k</sub> encloses exactly the "expected" eigenvalues, almost surely for all large N.
- The integral gives the estimator

$$\hat{t}_k = \frac{n}{N_k} \sum_{m \in N_k} \left( \lambda_m - \mu_m \right)$$

with  $\mathcal{N}_k$  the indexes of cluster k and  $\mu_1 \leq \ldots \leq \mu_N$  are the ordered eigenvalues of the matrix  $\operatorname{diag}(\boldsymbol{\lambda}) - \frac{1}{n}\sqrt{\lambda}\sqrt{\lambda}^{\mathrm{T}}$ ,  $\boldsymbol{\lambda} = (\lambda_1, \ldots, \lambda_N)^{\mathrm{T}}$ .

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R. Couillet, J. W. Silverstein, Z. Bai, M. Debbah, "Eigen-Inference for Energy Estimation of Multiple Sources," IEEE Trans. on Inf. Theory, vol. 57, no. 4, pp. 2420-2439, 2011.

Extending Y with zeros, our model is a "double sample covariance matrix"

$$\underbrace{\mathbf{\underline{Y}}}_{(N+n)\times M} = \underbrace{\begin{bmatrix} \mathbf{HP}^{\frac{1}{2}} & \sigma \mathbf{I}_{N} \\ \mathbf{0} & \mathbf{0} \end{bmatrix}}_{(N+n)\times (N+n)} \underbrace{\begin{bmatrix} \mathbf{X} \\ \mathbf{W} \end{bmatrix}}_{(N+n)\times M}$$

• Limiting distribution of  $\frac{1}{M} \mathbf{Y} \mathbf{Y}^{H}$ 

#### Theorem (I.s.d. of $\mathbf{B}_N$ )

Let  $\mathbf{B}_N = \frac{1}{M} \mathbf{Y} \mathbf{Y}^H$  with eigenvalues  $\lambda_1, \ldots, \lambda_N$ . Denote  $m_{\underline{\mathbf{B}}_N}(z) \stackrel{\Delta}{=} \frac{1}{M} \sum_{k=1}^M \frac{1}{\lambda_k - z}$ , with  $\lambda_i = 0$  for i > N. Then, for  $M/N \to c$ ,  $N/n_k \to c_k$ ,  $N/n \to c_0$ , for any  $z \in \mathbb{C}^+$ ,

$$m_{\underline{\mathbf{B}}_N}(z) \xrightarrow{\text{a.s.}} m_{\underline{F}}(z)$$

with  $m_F(z)$  the unique solution in  $\mathbb{C}^+$  of

$$\frac{1}{m_{\underline{F}}(z)} = -\sigma^2 + \frac{1}{f(z)} \left[ \frac{c_0 - 1}{c_0} + m_P\left(-\frac{1}{f(z)}\right) \right], \text{ with } f(z) = (c - 1)m_{\underline{F}}(z) - czm_{\underline{F}}(z)^2.$$

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estimator calculus

# Theorem (Estimator of $P_1, \ldots, P_K$ )

Let  $\mathbf{B}_N \in \mathbb{C}^{N \times N}$  be defined as above and  $\lambda = (\lambda_1, \ldots, \lambda_N)$ ,  $\lambda_1 < \ldots < \lambda_N$ . Assume that asymptotic cluster separability condition is fulfilled for some k. Then, as N, n,  $M \to \infty$ ,

$$\hat{P}_k - P_k \xrightarrow{\text{a.s.}} 0,$$

where

$$\hat{\boldsymbol{P}}_{k} = \frac{NM}{n_{k}(M-N)} \sum_{i \in \mathcal{N}_{k}} (\eta_{i} - \mu_{i})$$

with  $\mathcal{N}_k$  the set indexing the eigenvalues in cluster k of F,  $\eta_1 < \ldots < \eta_N$  the eigenvalues of diag $(\lambda) - \frac{1}{N}\sqrt{\lambda}\sqrt{\lambda}^T$  and  $\mu_1 < \ldots < \mu_N$  the eigenvalues of diag $(\lambda) - \frac{1}{M}\sqrt{\lambda}\sqrt{\lambda}^T$ .

## • solution is computationally simple, explicit, and the final formula compact.

- cluster separability condition is fundamental. This requires
  - for all other parameters fixed, the Pk cannot be too close to one another: source separation problem.
  - for all other parameters fixed,  $\sigma^2$  must be kept low: low SNR undecidability problem.
  - for all other parameters fixed, M/N cannot be too low: sample deficiency issue (not such an issue though).
  - for all other parameters fixed, *N*/*n* cannot be too low: diversity issue.
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Figure: Distribution function for the detection of two power sources,  $P_1 = 1$ ,  $P_2 = 4$ ,  $n_1 = n_2 = 1$ , M = N = 16. Optimum against Stieltjes transform method.

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Figure: Histogram of the cluster-mean approach and of  $\hat{P}_k$  for  $k \in \{1, 2, 3\}$ ,  $P_1 = 1/16$ ,  $P_2 = 1/4$ ,  $P_3 = 1$ ,  $n_1 = n_2 = n_3 = 4$  antennas per user, N = 24 sensors, M = 128 samples and SNR = 20 dB.





Figure: Normalized mean square error of largest estimated power  $\hat{P}_3$ ,  $P_1 = 1/16$ ,  $P_2 = 1/4$ ,  $P_3 = 1$ ,  $n_1 = n_2 = n_3 = 4$ , N = 24, M = 128. Comparison between classical, moment and Stieltjes transform approaches.

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#### Tools for Random Matrix Theory

- Classical Random Matrix Theory
- Introduction to Large Dimensional Random Matrix Theory
- The Random Matrix Pioneers
- The Moment Approach and Free Probability
- Introduction of the Stieltjes Transform
- Properties of the Asymptotic Support and Spiked Models
- Summary of what we know and what is left to be done

## 2 Random Matrix Theory and Signal Source Sensing

- Small Dimensional Analysis
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- Random matrix models of local failures in sensor networks
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# Failure detection



Random Matrix Theory and Failure Detection in Complex Systems Random matrix models of local failures in sensor networks Node failure detection in sensor networks

Consider the model

 $\mathbf{y} = \mathbf{H}\boldsymbol{\theta} + \sigma \mathbf{w}$ 

with  $\mathbf{H} \in \mathbb{C}^{N \times p}$  deterministic,  $\boldsymbol{\theta} \sim \mathcal{CN}(0, \mathbf{I}_p)$ ,  $\mathbf{w} \sim \mathcal{CN}(0, \mathbf{I}_N)$ .

- In particular  $E[\mathbf{y}] = \mathbf{0}$  and  $E[\mathbf{y}\mathbf{y}^{H}] = \mathbf{R} \stackrel{\Delta}{=} \mathbf{H}\mathbf{H}^{H} + \sigma^{2}\mathbf{I}_{N}$
- With  $\mathbf{s} = \mathbf{R}^{-\frac{1}{2}}\mathbf{y}$ ,

$$E[ss^H] = I_N.$$

• Upon failure of sensor k, y becomes

$$\mathbf{y}' = (\mathbf{I}_N - \mathbf{e}_k \mathbf{e}_k^{\mathsf{H}})\mathbf{H}\boldsymbol{\theta} + \sigma_k \mathbf{e}_k \mathbf{e}_k^* \boldsymbol{\theta}' + \sigma \mathbf{w}$$

for some noise variance  $\sigma_k^2$ .

Now E[y'] = 0 and

$$\mathbf{E}[\mathbf{y}'\mathbf{y}'^{\mathsf{H}}] = (\mathbf{I}_{N} - \mathbf{e}_{k}\mathbf{e}_{k}^{\mathsf{H}})\mathbf{H}\mathbf{H}^{\mathsf{H}}(\mathbf{I}_{N} - \mathbf{e}_{k}\mathbf{e}_{k}^{\mathsf{H}}) + \sigma_{k}^{2}\mathbf{e}_{k}\mathbf{e}_{k}^{\mathsf{H}} + \sigma^{2}\mathbf{I}_{N}$$

• With now  $\mathbf{s} = \mathbf{R}^{-\frac{1}{2}} \mathbf{y}'$ ,

$$\mathbf{E}[\mathbf{s}\mathbf{s}^{\mathsf{H}}] = \mathbf{I}_N + \mathbf{P}_k$$

with

$$\mathbf{P}_k = -\mathbf{R}^{-\frac{1}{2}}\mathbf{H}\mathbf{H}^{\mathsf{H}}\mathbf{e}_k\mathbf{e}_k^{\mathsf{H}}\mathbf{R}^{-\frac{1}{2}} + \mathbf{R}^{-\frac{1}{2}}\mathbf{e}_k\left[(\mathbf{e}_k^{\mathsf{H}}\mathbf{H}\mathbf{H}^{\mathsf{H}}\mathbf{e}_k + \sigma_k^2)\mathbf{e}_k^{\mathsf{H}}\mathbf{R}^{-\frac{1}{2}} - \mathbf{e}_k^{\mathsf{H}}\mathbf{H}\mathbf{H}^{\mathsf{H}}\mathbf{R}^{-\frac{1}{2}}\right]$$

of rank-2 (image of  $P_k$  in Span $(\mathbf{R}^{-\frac{1}{2}}\mathbf{e}_k, \mathbf{R}^{-\frac{1}{2}}\mathbf{HH}^{\mathsf{H}}\mathbf{e}_k)$ 

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with

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of rank-2 (image of  $\mathbf{P}_k$  in  $\text{Span}(\mathbf{R}^{-\frac{1}{2}}\mathbf{e}_k, \mathbf{R}^{-\frac{1}{2}}\mathbf{H}\mathbf{H}^{\mathsf{H}}\mathbf{e}_k))$ 

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• Upon sudden change of parameter  $\theta_k$ ,

$$\mathbf{y}' = \mathbf{H}(\mathbf{I}_{p} + \alpha_{k}\mathbf{e}_{k}\mathbf{e}_{k}^{*})\boldsymbol{\theta} + \mu_{k}\mathbf{H}\mathbf{e}_{k} + \sigma\mathbf{w}$$

Then

$$\mathbb{E}[\mathbf{y}'\mathbf{y}'^{\mathsf{H}}] = \mathbf{H}(\mathbf{I}_{\rho} + [\mu_{k}^{2} + (1 + \alpha_{k})^{2} - 1]\mathbf{e}_{k}\mathbf{e}_{k}^{\mathsf{H}})\mathbf{H}^{\mathsf{H}} + \sigma^{2}\mathbf{I}_{N}.$$

• With  $\mathbf{R} = \mathbf{H}\mathbf{H}^{\mathsf{H}} + \sigma^2 \mathbf{I}_N$  and  $\mathbf{s} = \mathbf{R}^{-\frac{1}{2}} \mathbf{y}'$ ,

$$\mathbf{E}[\mathbf{s}\mathbf{s}^{\mathsf{H}}] = \mathbf{I}_{N} + \mathbf{P}_{k}$$

with

$$\mathbf{P}_{k} = [\mu_{k}^{2} + (1 + \alpha_{k})^{2} - 1]\mathbf{R}^{-\frac{1}{2}}\mathbf{H}\mathbf{e}_{k}\mathbf{e}_{k}^{\mathsf{H}}\mathbf{H}^{\mathsf{H}}\mathbf{R}^{-\frac{1}{2}}.$$

#### Tools for Random Matrix Theory

- Classical Random Matrix Theory
- Introduction to Large Dimensional Random Matrix Theory
- The Random Matrix Pioneers
- The Moment Approach and Free Probability
- Introduction of the Stieltjes Transform
- Properties of the Asymptotic Support and Spiked Models
- Summary of what we know and what is left to be done

## 2 Random Matrix Theory and Signal Source Sensing

- Small Dimensional Analysis
- Large Dimensional Random Matrix Analysis

#### 3 Random Matrix Theory and Multi-Source Power Estimation

- Optimal detector
- The moment method
- The Stieltjes transform method

## Bandom Matrix Theory and Failure Detection in Complex Systems

- Random matrix models of local failures in sensor networks
- Failure detection and localization

• With K the number of failure scenarios, hypothesis test between:

- no failure
- failure of type 1
- ...
- failure of type K

Maximum-likelihood approach computationally constraining!

calculus cost  $\simeq O(N^3 K)$ 

which is

# calculus cost $\simeq O(N^{3+m})$

for *m* simultaneous node failures detection.

- Ad-hoc approaches/PCA can reduce this amount
- We propose here a "maximum-likelihood-type" method in

one SVD + O(K)

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R. Couillet and W. Hachem, "Local failure detection and identification in large sensor networks," *submitted to* IEEE Transaction on Information Theory, 2011.

- Upon reception of  $\mathbf{S} = [\mathbf{s}_1, \dots, \mathbf{s}_n]$ ,
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      <sub>0</sub>: failure
  - If  $\bar{\mathcal{H}}_0$  is decided, multi-hypothesis test

$$\mathcal{H}_k =$$
 "failure of type k"

• Detection test on largest eigenvalue  $\hat{\lambda}_1$  of  $\frac{1}{n}SS^H$ : for a false alarm rate  $\eta$ ,

$$\hat{\lambda}_1' \underset{\bar{\mathcal{H}}_0}{\overset{\mathcal{H}_0}{\lessgtr}} (T_2)^{-1} (1-\eta)$$

with

$$\hat{\lambda}_{1}' = N^{\frac{2}{3}} \frac{\hat{\lambda}_{1} - (1 + \sqrt{c_{N}})^{2}}{(1 + \sqrt{c_{N}})^{\frac{4}{3}} c_{N}^{\frac{1}{2}}}$$

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- For localization, eigenvalues are poor statistics
- Denote, in case of failure of type k

$$\mathbf{E}[\mathbf{s}\mathbf{s}^{\mathsf{H}}] = \mathbf{I}_{N} + \omega_{k}\mathbf{u}_{k,1}\mathbf{u}_{k,1}^{\mathsf{H}}$$

• We use the eigenvector  $\hat{\mathbf{u}}_1$  corresponding to  $\lambda_1$ , and

$$|\hat{\mathbf{u}}_1^{\mathsf{H}}\mathbf{u}_{k,1}|^2 \xrightarrow{\text{a.s.}} \xi(\omega_k) > 0$$

for *k* the failure index.

• With the CLT on  $|\hat{\mathbf{u}}_1^{\mathsf{H}}\mathbf{u}_{k,1}|^2 - \xi(\omega_k)$ , we have the estimator

$$k^* = \arg \max_{1 \le k \le K} f\left(\sqrt{N}(|\hat{\mathbf{u}}_1^{\mathsf{H}} \mathbf{u}_{k,1}|^2 - \xi(\omega_k)); \sigma_k^2\right)$$

with f the Gaussian density.

- Test can be reinforced by including
  - projection statistics on other vectors
  - statistics of eigenvalues
  - take the joint probability over multiple spikes.
- Further generalizations are possible assuming unknown failure amplitude.

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Figure: Correct detection (CDR) and localization (CLR) rates for different false alarm rates (FAR) and different *n*, worst case node failure in a 100-node network.

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#### Articles in Journals,

- R. Couillet, W. Hachem, "Local failure detection and identification in large sensor networks," IEEE Transactions on Information Theory, *submitted*.
- R. Couillet, J. Hoydis, M. Debbah, "Random Unitary Beamforming over Correlated Fading Channels," IEEE Transactions on Information Theory, *submitted*.
- R. Couillet, J. Hoydis, M. Debbah, "A deterministic equivalent approach to the performance analysis of isometric random precoded systems," IEEE Transactions on Information Theory, *submitted*.
- R. Couillet, J. W. Silverstein, Z. Bai, M. Debbah, "Eigen-Inference for Energy Estimation of Multiple Sources," IEEE Trans. on Information Theory, 2010, *to be published*.
- R. Couillet, J. W. Silverstein, M. Debbah, "A Deterministic Equivalent for the Capacity Analysis of Correlated Multi-User MIMO Channels," IEEE Trans. on Information Theory, *to be published*.
- P. Bianchi, J. Najim, M. Maida, M. Debbah, "Performance of Some Eigen-based Hypothesis Tests for Collaborative Sensing," IEEE Trans. on Information Theory, *to be published*.
- R. Couillet, M. Debbah, "A Bayesian Framework for Collaborative Multi-Source Signal Sensing," IEEE Trans. on Signal Processing, vol. 58, no. 10, pp. 5186-5195, 2010.
- S. Wagner, Ř. Couillet, M. Debbah, D. Slock, "Large System Analysis of Linear Precoding in MISO Broadcast Channels with Limited Feedback," IEEE Trans. on Information Theory, 2010, submitted.
- A. Masucci, Ø. Ryan, S. Yang, M. Debbah, "Gaussian Finite Dimensional Statistical Inference," IEEE Trans. on Information Theory, 2009, *submitted*.
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- M. Debbah, R. Müller, "MIMO channel modeling and the principle of maximum entropy," IEEE Trans. on Information Theory, vol. 51, no. 5, pp. 1667-1690, 2005.

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#### Articles in International Conferences

- A. Kammoun, R. Couillet, J. Najim, M. Debbah, "A G-estimator for rate adaption in cognitive radios," *submitted to* IEEE International Symposium on Information Theory, St Petersburg, Russia, 2011.
- J. Yao, R. Couillet, J. Najim, E. Moulines, M. Debbah, "CLT for eigen-inference methods in cognitive radios," IEEE International Conf. on Acoustics, Speech and Signal Proc., Prague, Czech Rep., 2011.
- J. Hoydis, R. Couillet, M. Debbah, "Deterministic Equivalents for the Performance Analysis of Isometric Random Precoded Systems," IEEE International Conference on Communications, Kyoto, Japan, 2011.
- J. Hoydis, J. Najim, R. Couillet, M. Debbah, "Fluctuations of the Mutual Information in Large Distributed Antenna Systems with Colored Noise," Forty-Eighth Annual Allerton Conference on Communication, Control, and Computing, Allerton, IL, USA, 2010.
- R. Couillet, S. Wagner, M. Debbah, A. Silva, "The Space Frontier: Physical Limits of Multiple Antenna Information Transfer", Inter-Perf 2008, Athens, Greece. **BEST STUDENT PAPER AWARD.**
- R. Couillet, M. Debbah, V. Poor, "Self-organized spectrum sharing in large MIMO multiple access channels", submitted to ISIT 2010.
- L. S. Cardoso, M. Debbah, P. Bianchi, and J. Najim, "Cooperative spectrum sensing using random matrix theory," 3rd International Symposium on Wireless Pervasive Computing (ISWPC), 2008.
- R. Couillet, M. Debbah, "Uplink capacity of self-organizing clustered orthogonal CDMA networks in flat fading channels", ITW 2009 Fall, Taormina, Sicily.

## Book Chapters

#### Mathematical Foundations for Signal Processing, Communications and Networking

- Editors: T. Chen, D. Rajan and E. Serpedin
- Chapter title: "Random matrix theory"
- Chapter authors: R. Couillet and M. Debbah
- Publisher: CRC Press, Taylor & and Francis Group
- Year: 2011 (to appear)

# Coming up soon...



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#### Romain Couillet, Mérouane Debbah, Random Matrix Methods for Wireless Communications.

# Theoretical aspects

- Random matrices
- 2 The Stieltjes transform method
- 3 Free probability theory
- ④ Combinatoric approaches
- 6 Deterministic equivalents
- 6 Spectrum analysis
- Ø Eigen-inference
- 8 Extreme eigenvalues
- Summary and partial conclusions
- 2 Applications to wireless communications
  - Introduction to applications in telecommunications
  - Ø System performance of CDMA technologies
  - ③ Performance of multiple antennas systems
  - 4 Rate performance in multiple access and broadcast channels
  - 9 Performance of multi-cellular and relay networks
  - 6 Detection
  - Ø Estimation
  - 8 System modelling
  - 9 Perspectives
  - Occurrence Conclusion

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