# Digital Communications Exercises 

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## Pulse amplitude modulation

## Exercise 1

Consider a real channel $Y=X+N$, with $N \sim \mathcal{N}\left(0, N_{0}\right), N_{0}=0.1$, and suppose that $\mathrm{E}\left[X^{2}\right] \leq 1$. If the input lies in an $L$-ary alphabet $A=\{ \pm 1 c, \pm 3 c, \ldots, \pm(L-1) c\}$, how big can be $L$ so that the probability of error is about $10^{-6}$ using ML detection?

## Pulse amplitude modulation

## Exercise 1

Consider the probability of error $P(e)$ for PAM modulation

$$
P(e)=\frac{2(L-1)}{L} Q\left(\sqrt{\frac{6 \mathrm{E}\left[X^{2}\right]}{\left(L^{2}-1\right) N_{0}}}\right)
$$

and determine the largest possible $P(e)$ (using possibly the approximation of the $Q$ function.

## MAP detection

## Exercise 2

A discrete communication channel with input $x$ and output $y$ has input alphabet $\{0,1\}$, output alphabet $\{a, b, c, d\}$ and transition probability $P(y \mid x)$ given by

$$
\begin{align*}
& P(a \mid 0)=0.3,  \tag{1}\\
& P(a \mid 1)=0.1,  \tag{2}\\
& P(b \mid 0)=0.1, \\
& P(b \mid 1)=0.5,  \tag{4}\\
& P(c \mid 0)=0.5,  \tag{5}\\
& P(c \mid 1)=0.1,  \tag{6}\\
& P(d \mid 0)=0.1,  \tag{7}\\
& P(d \mid 1)=0.3 . \tag{8}
\end{align*}
$$

The a priori probability of the input is $P(x=0)=0.6$ and $P(x=1)=0.4$. Compute the MAP detection rule and the resulting average error probability.

## MAP detection

## Exercise 2

The decision region for 0 is the set of outputs (subset of $\{a, b, c, d\}$ ), which satisfies

$$
P(0 \mid y)>P(1 \mid y)
$$

Using Bayes' rule, this is the set of $y$ 's that satisfies

$$
P(y \mid 0) P(0)>P(y \mid 1) P(1) .
$$

Call $D_{0}$ the detection region for 0 and $D_{1}$ the detection region for 1 . The probability of error is then

$$
P(e)=P(e \mid 0) P(0)+P(e \mid 1) P(1)=P\left(y \in D_{1} \mid 0\right) P(0)+P\left(y \in D_{0} \mid 1\right) P(1) .
$$

## Laplacian Noise

## Exercise 3

Consider a real binary input Laplacian noise channel $y=x+n$, with input alphabet $\{-1,+A\}$ and noise pdf

$$
p(n)=\frac{b}{2} e^{-b|n|} .
$$

The prior input probability distribution is $P(A)=0.8$ and $P(-A)=0.2$. Find the MAP decision regions and compute the average decision error probability. Compare to the Gaussian case.

## Laplacian Noise

## Exercise 3

We proceed similarly as for Exercise 2. Now the error probability is continuous and therefore $P\left(y \in D_{-A} \mid+A\right)$ and $P\left(y \in D_{+A} \mid-A\right)$ are integral forms.
Solution: Two cases must be isolated, depending on whether $A<\log (2) / b$. If so, $\mathcal{D}_{A}=\mathbb{R}$, otherwise, $\mathcal{D}_{A}=(-\log (2) / b, \infty)$.

## Constellations

## Exercise 4

Consider the signal constellation defined by the points

$$
\begin{align*}
& s_{m}=\sqrt{E_{1}} \exp (i(m-1) / 2+\pi / 4), \text { for } m=1, \ldots, 4  \tag{9}\\
& s_{m}=\sqrt{E_{2}} \exp (i(m-5) / 2), \text { for } m=5, \ldots, 8 \tag{10}
\end{align*}
$$

(1) Find the average energy per symbol $E_{s}$ (with equiprobabable entries) as a function of $E_{1}$ and $E_{2}$.
(2) For fixed $E_{s}$, find the ratio $\rho=E_{2} / E_{1}$ that maximizes the minimum squared Euclidean distance of the constellation (assume $E_{2} \geq E_{1}$ ).
(3) For the optimal ratio $\rho$ found, determine an upperbound to the symbol error probability as a function of $E_{b} / N_{0}$.

## Constellations

## Exercise 4

(1) $E_{S}$ is defined by

$$
E_{s}=\frac{1}{M} \sum_{m=1}^{M}\left|s_{m}\right|^{2}
$$

(2) The distance $d_{\min }$ between two close points is given by

$$
d_{\min }=\left\|s_{1}-s_{5}\right\| .
$$

It suffices to write $d_{\text {min }}$ as a function of $E_{1}$ alone, find $E_{1}$ that reaches the minimum, and determine $\rho$.
(3) Use the fact that error regions for points further than the two points at $d_{\text {min }}$ are included in the error regions for the closer points. Hence

$$
P(e) \leq 2 Q\left(\sqrt{\frac{\log _{2}(M) d_{\min }^{2}}{2} \frac{E_{b}}{N_{0}}}\right)
$$

## PSK modulation

## Exercise 5

In a 4-QAM modulator, each symbol is labeled by the pair of bits $\left(b_{1} b_{2}\right)$ according to the following labeling rule:

$$
\begin{align*}
\mathbf{a}_{0}=\sqrt{E} e^{j \pi / 4} & \mapsto 00,  \tag{11}\\
\mathbf{a}_{1}=\sqrt{E} e^{3 j \pi / 4} & \mapsto 01,  \tag{12}\\
\mathbf{a}_{2}=\sqrt{E} e^{-3 j \pi / 4} & \mapsto 10,  \tag{13}\\
\mathbf{a}_{3}=\sqrt{E} e^{-j \pi / 4} & \mapsto 11 . \tag{14}
\end{align*}
$$

The bits $b_{1}$ and $b_{2}$ are statistically independent with probability $P\left(b_{i}=0\right)=\varepsilon, P\left(b_{1}=1\right)=1-\varepsilon$, $\varepsilon<0.5$. The signal is modulated and transmitted over an AWGN channel with power spectral density $N_{0}$.
Determine the decision regions of the MAP detector (that minimizes the symbol error probability) and the decision regions of a detector that minimizes the bit error probability.

## PSK modulation

## Exercise 5

The MAP decision regions unfold as in previous exercises. As for the bit error probability, consider individually the first and second bit. The probability of bit error is the average.

## PSK modulation

## Exercise 6

Consider the input equiprobable sequence $a_{n} \in\{-1,+1\}$, that passes through a convolution channel $h_{n}$ with $H(z)=1+\frac{1}{2} z^{-1}$. The signal is affected by white Gaussian noise, and is received as

$$
y_{0}=1.2, y_{1}=-0.3, y_{2}=-0.5, y_{3}=0.1, y_{4}=1.5
$$

Assuming all sequences of $a_{0}, \ldots, a_{4}$ could be transmitted, what is the sequence detected by a symbol-by-symbol detector? What is the sequence detected by a Viterbi detector?

## PSK modulation

## Exercise 6

Symbol by symbol detection takes merely the sign function as a decision rule: $\hat{a}_{n}=\operatorname{sign}\left(y_{n}\right)$. Viterbi detection requires to design the state machine where the states are based on the four inputs $\left(a_{n}, a_{n-1}\right)$, but for the first stage where the states are the two inputs of $a_{0}$. The channel memory length here is $L=1$ and the correlator coefficients are $g_{0}=1, g_{1}=1 / 2$.

## ML detection

## Exercise 7

$x_{1}, \ldots, x_{N}$ are independent samples of a zero mean Gaussian random variable, whose variance is known. Give the expression of the estimated variance provided by ML estimation.

## ML detection

## Exercise 7

It suffices to find

$$
\arg \max _{\sigma^{2}} P\left(x_{1}, \ldots, x_{N} \mid \sigma^{2}\right) .
$$

This is done by remembering that $x_{1}, \ldots, x_{N}$ are independent, and therefore $P\left(x_{1}, \ldots, x_{N} \mid \sigma^{2}\right)=\prod_{i} P\left(x_{i} \mid \sigma^{2}\right)$. We then differentiate this expression, equate to zero and determine the maximizing estimate $\hat{\sigma^{2}}$ of $\sigma^{2}$.

## Non-uniform distribution

## Exercise 8

Consider a random variable $X \in\{-3 \alpha,-\alpha, \alpha, 3 \alpha\}$ with a priori probabilities $P( \pm \alpha)=0.4$, $P( \pm 3 \alpha)=0.1$. The parameter $\alpha$ is set so that the mean signal energy is 1 . Given an observation of $Y=X+N, N$ being zero mean real Gaussian with variance $\sigma^{2}$, independent on $X$, what are the MAP decision regions? If $\sigma^{2}=0.25$ and $Y=2.1$, what is the decision? What is the overall probability of error?

## Non-uniform distribution

## Exercise 8

Computing $\sum_{i=1}^{4} X_{i}^{2}=1$, we obtain $\alpha \simeq 0.62$. The decision regions are then given by

$$
\begin{aligned}
\mathcal{D}_{-3 \alpha} & =\left(-\infty, \frac{\sigma^{2}}{2 \alpha} \frac{\log (0.1)}{\log (0.4}-2 \alpha\right] \\
\mathcal{D}_{-\alpha} & =\left[\frac{\sigma^{2}}{2 \alpha} \frac{\log (0.1)}{\log (0.4}-2 \alpha, 0\right]
\end{aligned}
$$

the other regions being symmetric.

